

# Synchronous Sequential Circuit

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- ✓ The change of internal state occurs in response to the synchronized clock pulses.
- ✓ Data are read during the clock pulse (e.g. rising-edge triggered)
- ✓ It is supposed to wait long enough after the external input changes for all flip-flop inputs to reach a steady value before the next clock pulse
- ✓ Unsuitable Situations:
  - Inputs can change at any time and cannot be synchronized with a clock
  - Circuit is large, a cost in time of transitions can not be avoided

# Asynchronous Circuits

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- ✓ Not synchronized by a common clock
  - ✓ States change immediately after input changes
  - ✓ For a given value of input variables, the system is stable if the circuit reaches a steady state condition.
  - ✓ The circuit reaches a steady-state condition when  $y_i = Y_i$  for all  $i$ .
  - ✓ A transition from one stable state to another occurs only in response to a change in an input variable
- ✓ **Fundamental-mode operation**
    - The input signals change only when the circuit is in a **stable condition**
    - The input signals change **one at a time**
- ✓ The time between two input changes must be longer than the time it takes the circuit to reach a stable state.
  - ✓ Timing is a Major Problem because of **unequal delays** through various paths in the circuit

# Why Asynchronous Sequential Circuits?

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## Asynchronous sequential circuits basics

- ✓ No clock signal is required
- ✓ Internal states can change at any instant of time when there is a change in the input variables
- ✓ Have better performance but hard to design due to timing problems

## Why Asynchronous Circuits?

- ✓ Accelerate the speed of the machine (no need to wait for the next clock pulse).
- ✓ Simplify the circuit in the small independent gates.
- ✓ Necessary when having multi circuits each having its own clock.

## Analysis Procedure

- ✓ The analysis consists of obtaining a table or a diagram that describes the sequence of internal states and outputs as a function of changes in the input variables.

# Example Circuit

✓ First construction of Asynchronous Circuits:

- using only gates
- with **feedback paths**

✓ Analysis:

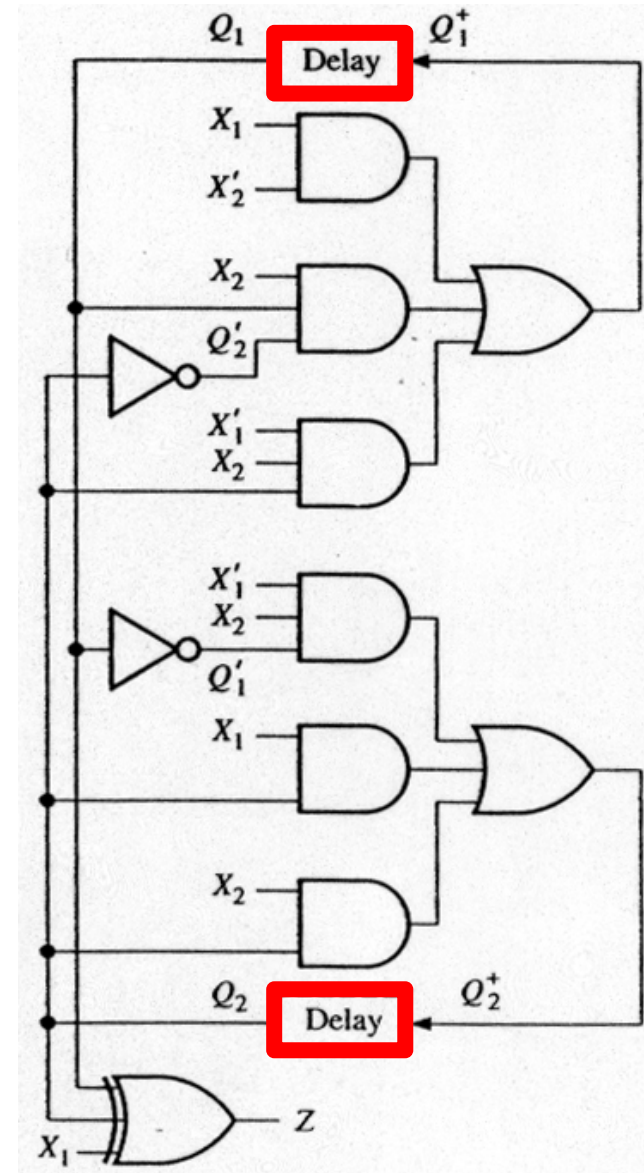
- Lump all of the delay associated with each feedback path into a "delay" box
- Associate a state variable with each delay output
- Construct the flow table

✓ Network equations

$$Q_1^+ = X_1 X_2' + X_1' X_2 Q_2 + X_2 Q_1 Q_2'$$

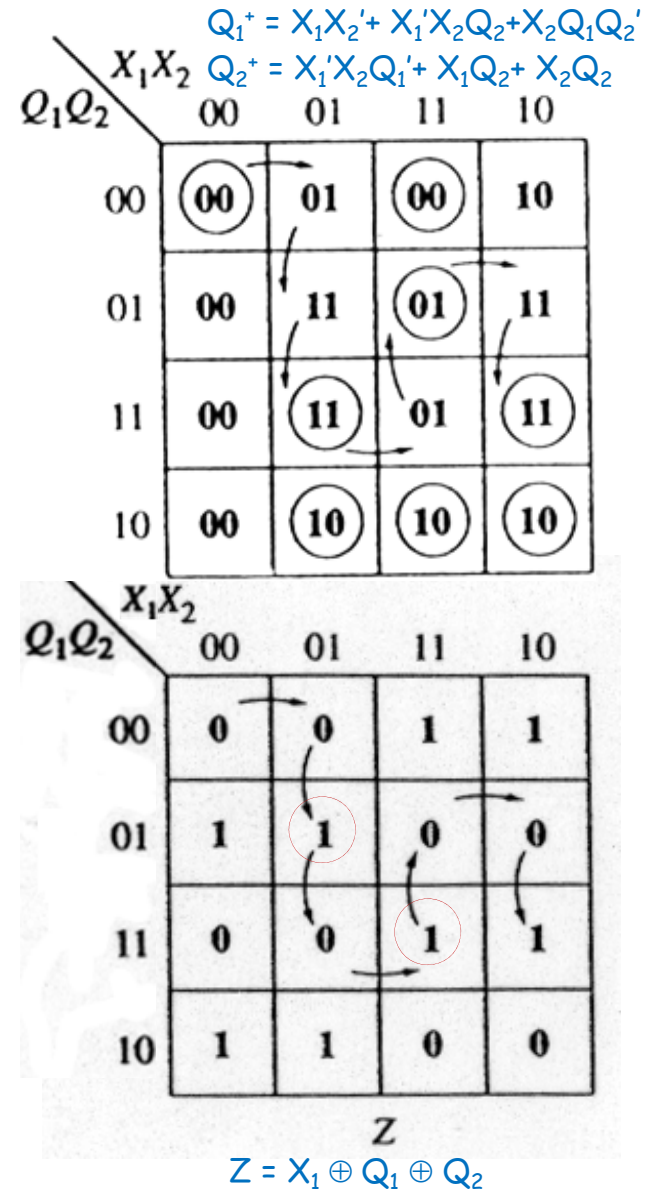
$$Q_2^+ = X_1' X_2 Q_1' + X_1 Q_2 + X_2 Q_2$$

$$Z = X_1 \oplus Q_1 \oplus Q_2$$



# Example Circuit: Output Table

- ✓ 1. Starting in total state  $X_1X_2Q_1Q_2=0000$
- ✓ 2. Input changes to 01
  - Internal state changes to 01 and then to 11.
- ✓ 3. Input changes to 11.
  - Go to unstable total state 1111 and then to 1101.
- ✓ 4. Input changes to 10.
  - Go to unstable total state 1001 and then to 1011.
- ✓ The output sequence:
  - 0 (0) (1) 0 (1) 0 (0) 1
  - Condensed to the form 0 (1) 0 (1) 0 1.
  - Two transient 1 outputs is dangerous can be eliminated by proper design.



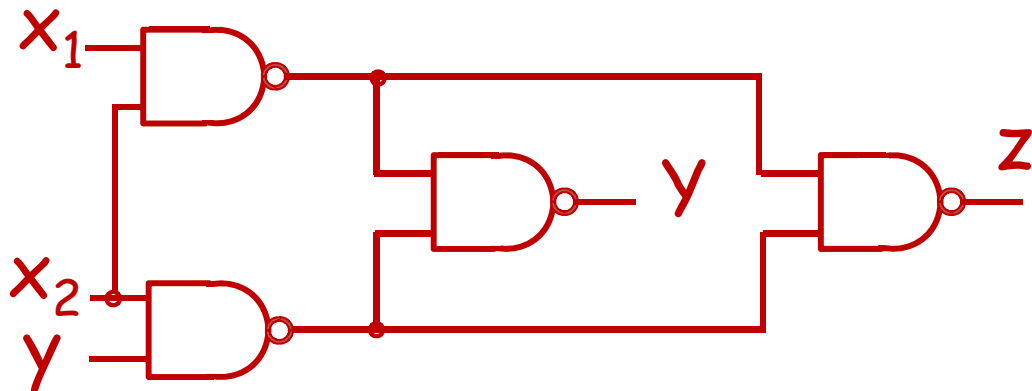
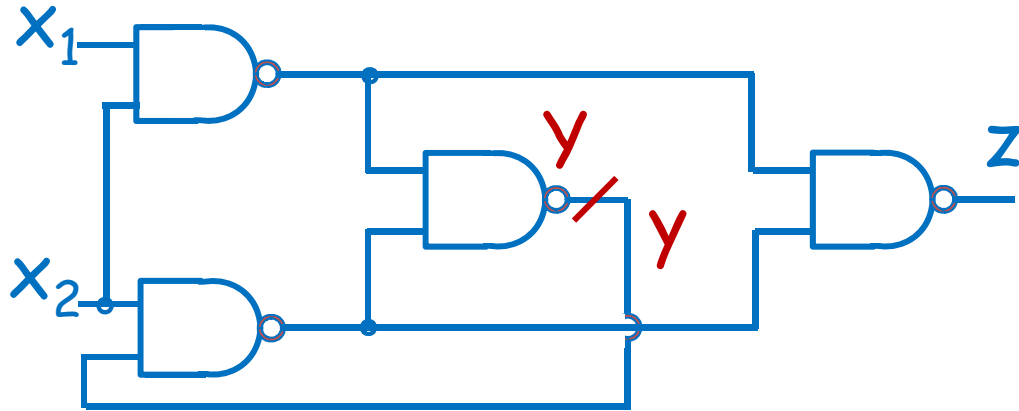
# Transition Table

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- ✓ **Transition table** is useful to analyze an asynchronous circuit from the circuit diagram. Procedure to obtain transition table:
  1. Determine all feedback loops in the circuits
  2. Mark the input ( $y_i$ ) and output ( $Y_i$ ) of each feedback loop
  3. Derive the Boolean functions of all  $Y$ 's
  4. Plot each  $Y$  function in a map and combine all maps into one table (**flow table**)
  5. Circle those values of  $Y$  in each square that are equal to the value of  $y$  in the same row (**stable states**)

# Asynchronous Sequential Analysis

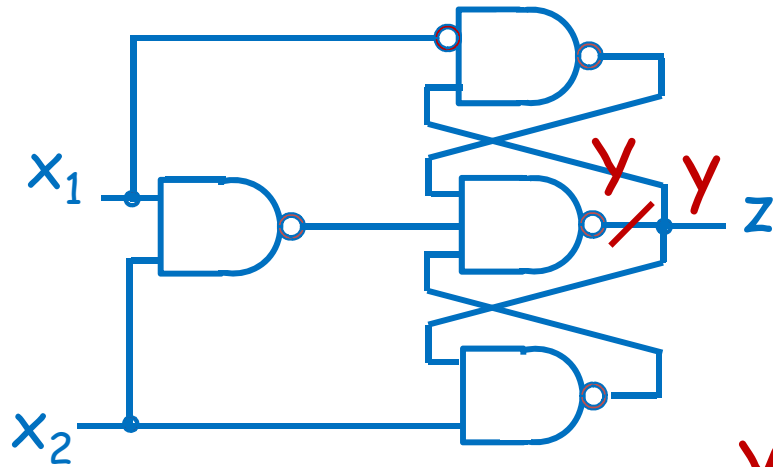
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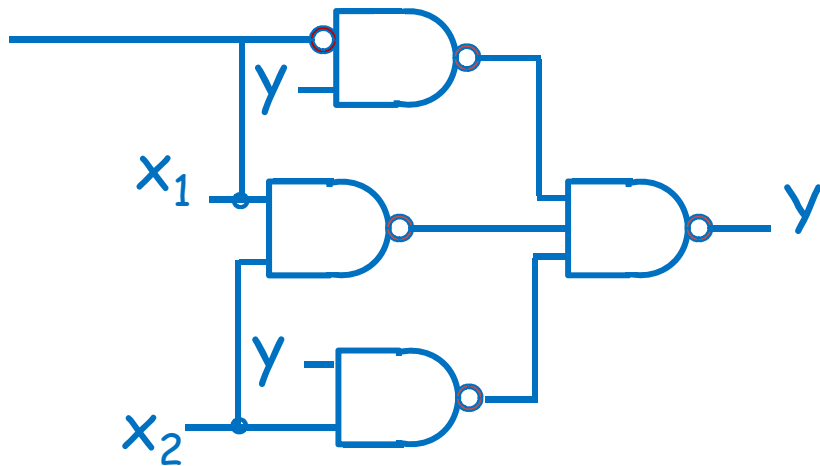
$$Y = \overline{\overline{x_1 x_2} \overline{x_2 y}} = x_1 x_2 + x_2 y$$
$$z = Y$$

# Asynchronous Sequential Analysis

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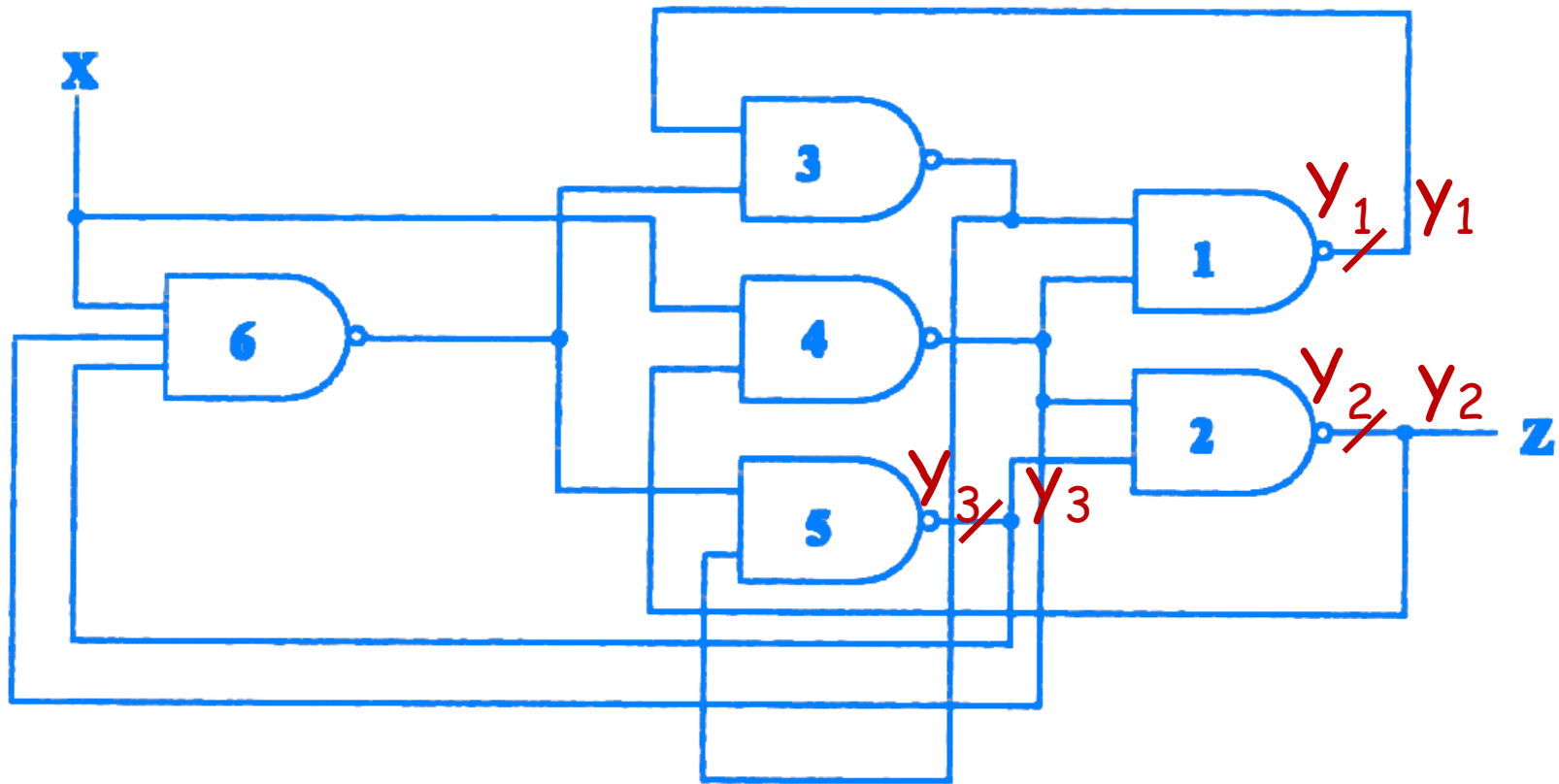
$$Y = \overline{x_1 x_2} \overline{x_2 y} \overline{y \overline{x_1}} = x_1 x_2 + x_2 y + y \overline{x_1}$$
$$z = y$$





# Asynchronous Sequential Analysis

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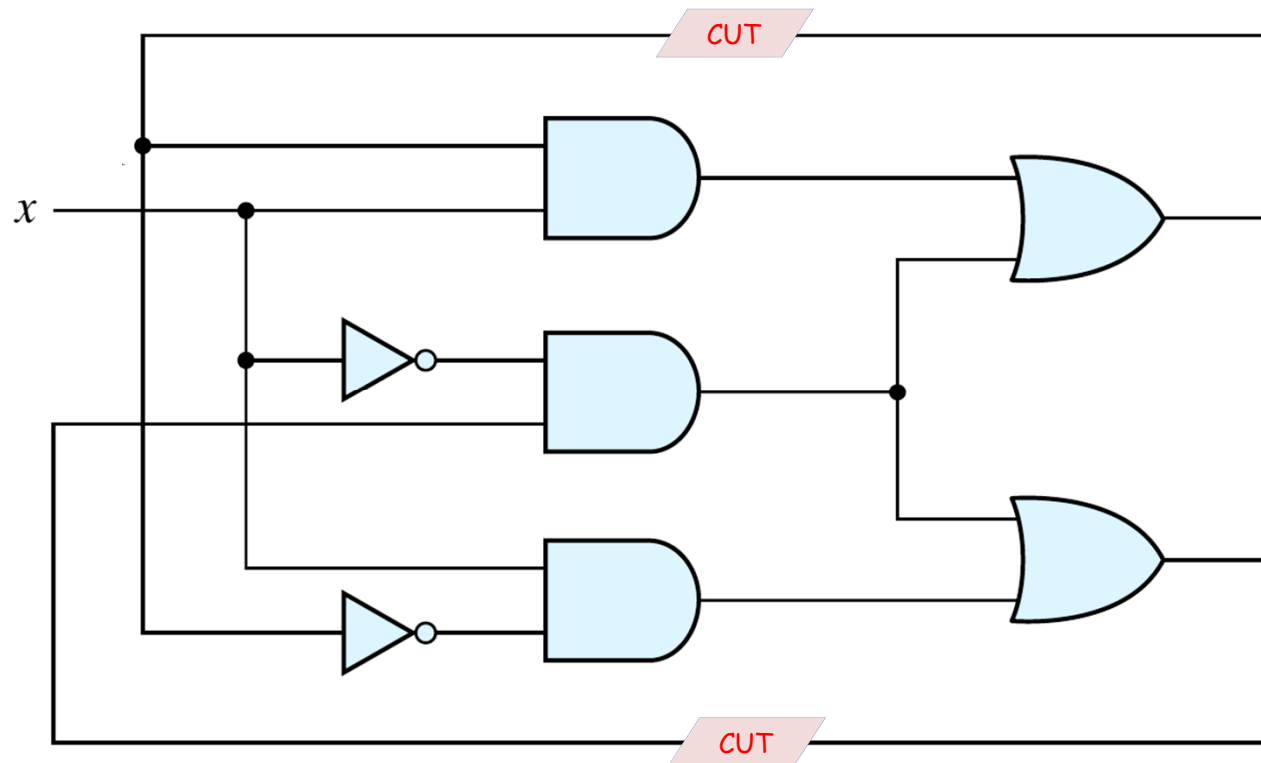


# Asynchronous Sequential Circuit

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✓ The state variables:  $Y_1$  and  $Y_2$

- $Y_1 = xy_1 + \bar{x}y_2$
- $Y_2 = x\bar{y}_1 + \bar{x}y_2$



# Transition Table

## ✓ Combine the internal state with input variables

- Stable total states:

$$y_1 y_2 x = 000, 011, 110 \text{ and } 101$$

$$Y_1 = xy_1 + \bar{x}y_2$$

$$Y_2 = x\bar{y}_1 + \bar{x}y_2$$

		$x$	
		0	1
$y_1 y_2$	00	0	0
	01	1	0
	11	1	1
	10	0	1

(a) Map for  
 $Y_1 = xy_1 + x'y_2$

		$x$	
		0	1
$y_1 y_2$	00	0	1
	01	1	1
	11	1	0
	10	0	0

(b) Map for  
 $Y_2 = xy'_1 + x'y_2$

		$x$	
		0	1
$y_1 y_2$	00	(00)	01
	01	11	(01)
	11	(11)	10
	10	00	(10)

# Transition Table

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- ✓ In an asynchronous sequential circuit, the internal state can change immediately after a change in the input.
- ✓ It is sometimes convenient to combine the internal state with input value together and call it the **Total State of the circuit**. (Total state = Internal state + Inputs)
- ✓ In the example , the circuit has
  - 4 stable total states: ( $y_1y_2x=$  000, 011, 110, and 101)
  - 4 unstable total states: ( $y_1y_2x=$  001, 010, 111, and 100)

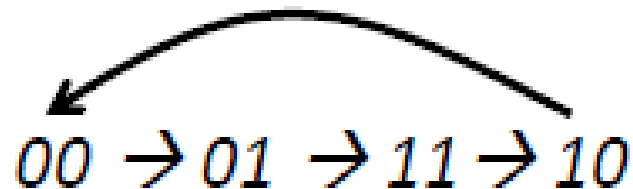
	$x$	
	0	1
$y_1y_2$		
00	00	01
01	11	01
11	11	10
10	00	10

# Transition Table

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- ✓ If  $y=00$  and  $x=0 \Rightarrow Y=00$  (Stable state)
- ✓ If  $x$  changes from 0 to 1 while  $y=00$ , the circuit changes  $Y$  to 01 which is temporary unstable condition ( $Y \neq y$ )
- ✓ As soon as the signal propagates to make  $Y=01$ , the feedback path causes a change in  $y$  to 01. (transition from the first row to the second row)
- ✓ If the input **alternates** between 0 and 1, the circuit will repeat the sequence of states:

	$x$	
	0	1
$y_1 y_2$		
00	00	01
01	11	01
11	11	10
10	00	10



# Flow Table

- ✓ A **flow table** is similar to a transition table except that the internal state are symbolized with letters rather than binary numbers.
- ✓ It also includes the output values of the circuit for each stable state.

$y \backslash x$	0	1
$a$	$\textcircled{a}$	$b$
$b$	$c$	$\textcircled{b}$
$c$	$\textcircled{c}$	$d$
$d$	$a$	$\textcircled{d}$

(a) Four states with one input

$\backslash x_1x_2$	00	01	11	10
$a$	$\textcircled{a}, 0$	$\textcircled{a}, 0$	$\textcircled{a}, 0$	$b, 0$
$b$	$a, 0$	$a, 0$	$\textcircled{b}, 1$	$\textcircled{b}, 0$

(b) Two states with two inputs and one output

# Flow Table

- ✓ In order to obtain the circuit described by a flow table, it is necessary to convert the flow table into a transition table from which we can derive the logic diagram.

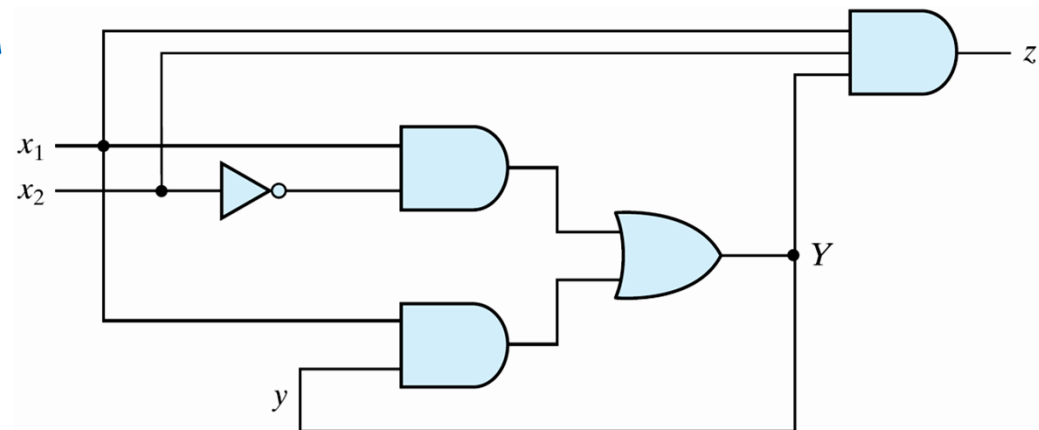
$x_1x_2$	00	01	11	10
a	(a, 0)	(a, 0)	(a, 0)	b, 0
b	a, 0	a, 0	(b, 1)	(b, 0)

Assignments:  
 $A \equiv 0$        $B \equiv 1$

$y$	$x_1x_2$	00	01	11	10
0	(0)	(0)	(0)	1	
1	0	0	(1)	(1)	

$y$	$x_1x_2$	00	01	11	10
0	0	0	0	0	
1	0	0	(1)	0	

- ✓ This can be done through the assignment of a distinct binary value to each state.



(c) Logic diagram

# Race condition

- ✓ Two or more binary state variables will change value when one input variable changes.
- ✓ Cannot predict state sequence if unequal delay is encountered.
- ✓ **Non-critical race:** The final stable state does not depend on the change order of state variables
- ✓ **Critical race:** The change order of state variables will result in different stable states. **Must be avoided !!**

	$x$	
	0	1
$y_1y_2$	00	11
	01	11
	11	11
	10	11

(a) Possible transitions:

00  $\rightarrow$  11  
 00  $\rightarrow$  01  $\rightarrow$  11  
 00  $\rightarrow$  10  $\rightarrow$  11

	$x$	
	0	1
$y_1y_2$	00	11
	01	01
	11	01
	10	11

(b) Possible transitions:

00  $\rightarrow$  11  $\rightarrow$  01  
 00  $\rightarrow$  01  
 00  $\rightarrow$  10  $\rightarrow$  11  $\rightarrow$  01

	$x$	
	0	1
$y_2$	00	11
	01	01
	11	11
	10	10

(a) Possible transitions:

00  $\rightarrow$  11  
 00  $\rightarrow$  01  
 00  $\rightarrow$  10

	$x$	
	0	1
$y_1y_2$	00	11
	01	11
	11	11
	10	10

(b) Possible transitions:

00  $\rightarrow$  11  
 00  $\rightarrow$  01  $\rightarrow$  11  
 00  $\rightarrow$  10



# Race Solution

- ✓ It can be solved by making a proper binary assignment to the state variables.
- ✓ The state variables must be assigned binary numbers in such a way that only one state variable can change at any one time when a state transition occurs in the flow table.

	$x$	0	1
$y_1y_2$	00	00	01
	01		11
	11		10
	10		10

(a) State transition:  
00 → 01 → 11 → 10

	$x$	0	1
$y_1y_2$	00	00	01
	01		11
	11		11
	10		10

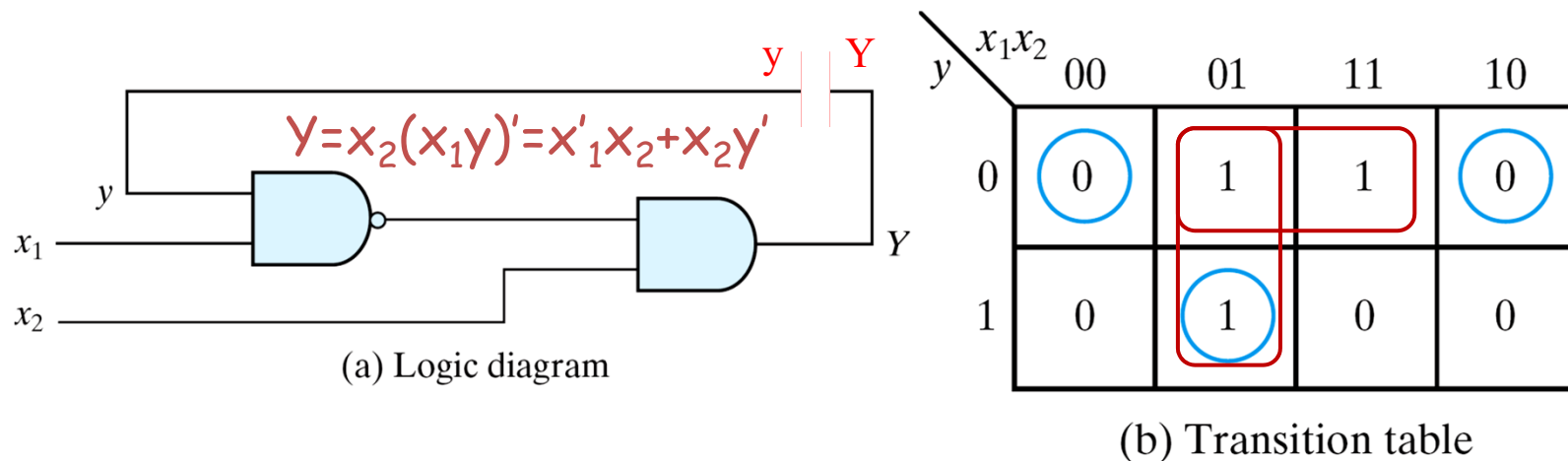
(b) State transition:  
00 → 01 → 11

	$x$	0	1
$y_1y_2$	00	00	01
	01		11
	11		10
	10		01

(c) Unstable  
01 → 11 → 10

# Stability Check

- ✓ Asynchronous sequential circuits may oscillate between unstable states due to the feedback
  - Must check for stability to ensure proper operations
- ✓ Can be easily checked from the transition table
  - Any column has no stable states  $\longrightarrow$  unstable
  - Ex: when  $x_1x_2=11$  in (b),  $Y$  and  $y$  are never the same

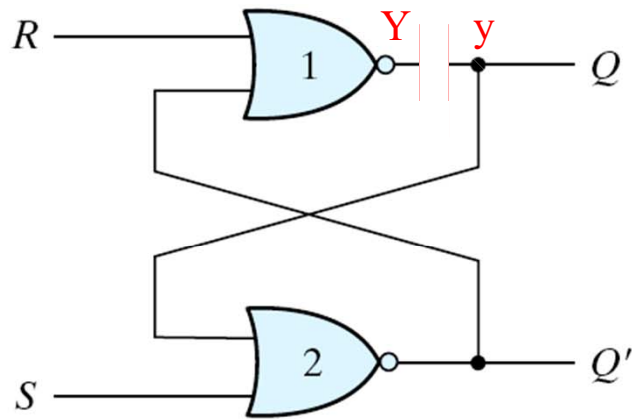


# Latches in Asynchronous Circuits

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- ✓ The traditional configuration of asynchronous circuits is using one or more feedback loops
  - No real delay elements.
- ✓ It is more convenient to employ the SR latch as a memory element in asynchronous circuits
  - Produce an orderly pattern in the logic diagram with the memory elements clearly visible.
- ✓ SR latch is an asynchronous circuit
  - So will be analyzed first using the method for asynchronous circuits.

# SR Latch with NOR Gates



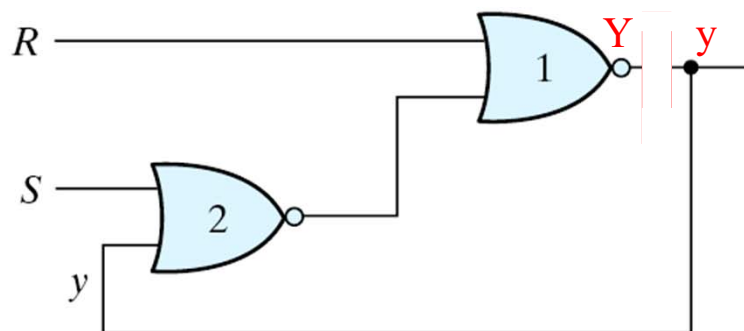
(a) Cross-coupled circuit

$S$	$R$	$Q$	$Q'$
1	0	1	0
0	0	1	0
0	1	0	1
0	0	0	1
1	1	0	0

(After  $SR = 10$ )

(After  $SR = 01$ )

(b) Truth table



(c) Circuit showing feedback

		$SR$			
		00	01	11	10
$y$	0	0	0	0	1
	1	1	0	0	1

$$Y = SR' + R'y$$

(d) Transition table

# Constraints on Inputs

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The condition to be avoided is that both S and R inputs must not be 1 simultaneously. This condition is avoided when  $SR = 0$  (i.e., ANDing of S and R must always result in 0).

When  $SR = 0$  holds at all times, the excitation function derived previously:

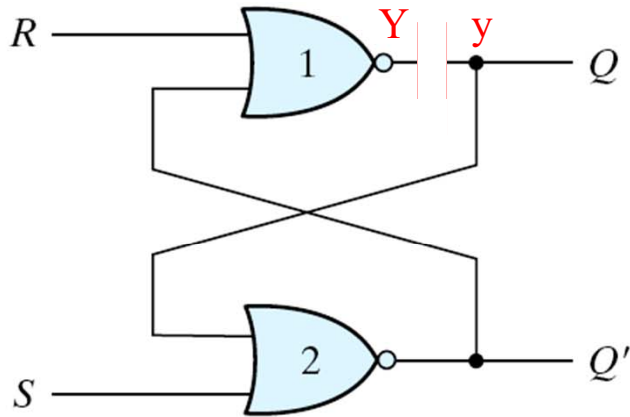
$$Y = SR' + R'y$$

can be expressed as:

$$Y = S + R'y$$

$y \backslash SR$	00	01	11	10
0	0	0	-	1
1	1	0	-	1

# SR Latch with NOR Gates



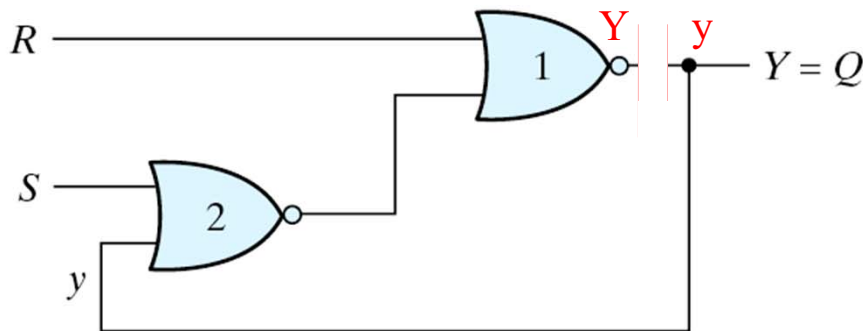
(a) Cross-coupled circuit

S	R	Q	Q'
1	0	1	0
0	0	1	0
0	1	0	1
0	0	0	1
1	1	0	0

(After  $SR = 10$ )

(After  $SR = 01$ )

(b) Truth table



(c) Circuit showing feedback

SR		y			
		00	01	11	10
y	0	0	0	0	1
	1	1	0	0	1

$$Y = SR' + R'y$$

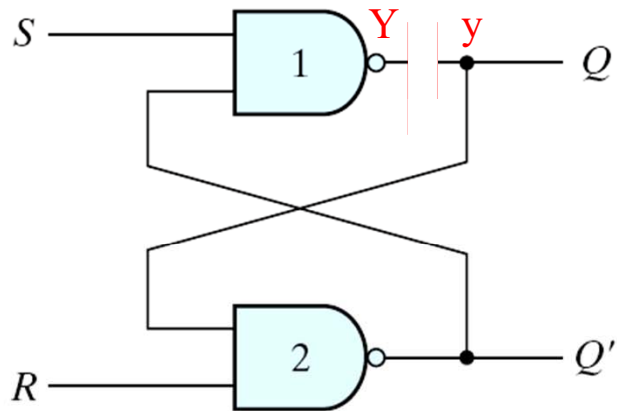
$$Y = S + R'y \text{ when } SR = 0 \rightarrow$$

(d) Transition table

S=1, R=1 ( $SR = 1$ )  
should not be used  
 $\Rightarrow SR = 0$  is  
normal mode

**should be  
carefully  
checked first**

# SR Latch with NAND Gates



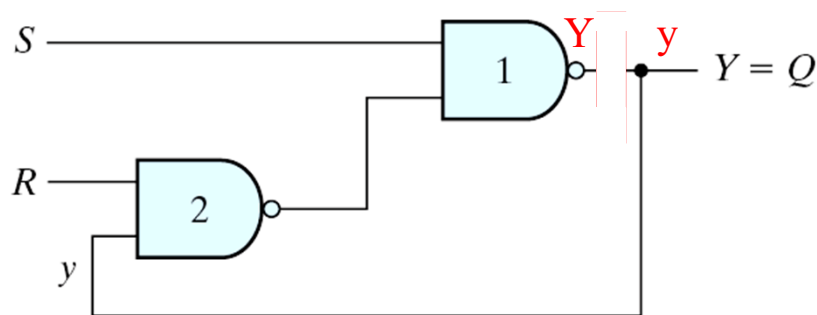
(a) Cross-coupled circuit

$S$	$R$	$Q$	$Q'$
1	0	0	1
1	1	0	1
0	1	1	0
1	1	1	0
0	0	1	1

(After  $SR = 10$ )

(After  $SR = 01$ )

(b) Truth table



(c) Circuit showing feedback

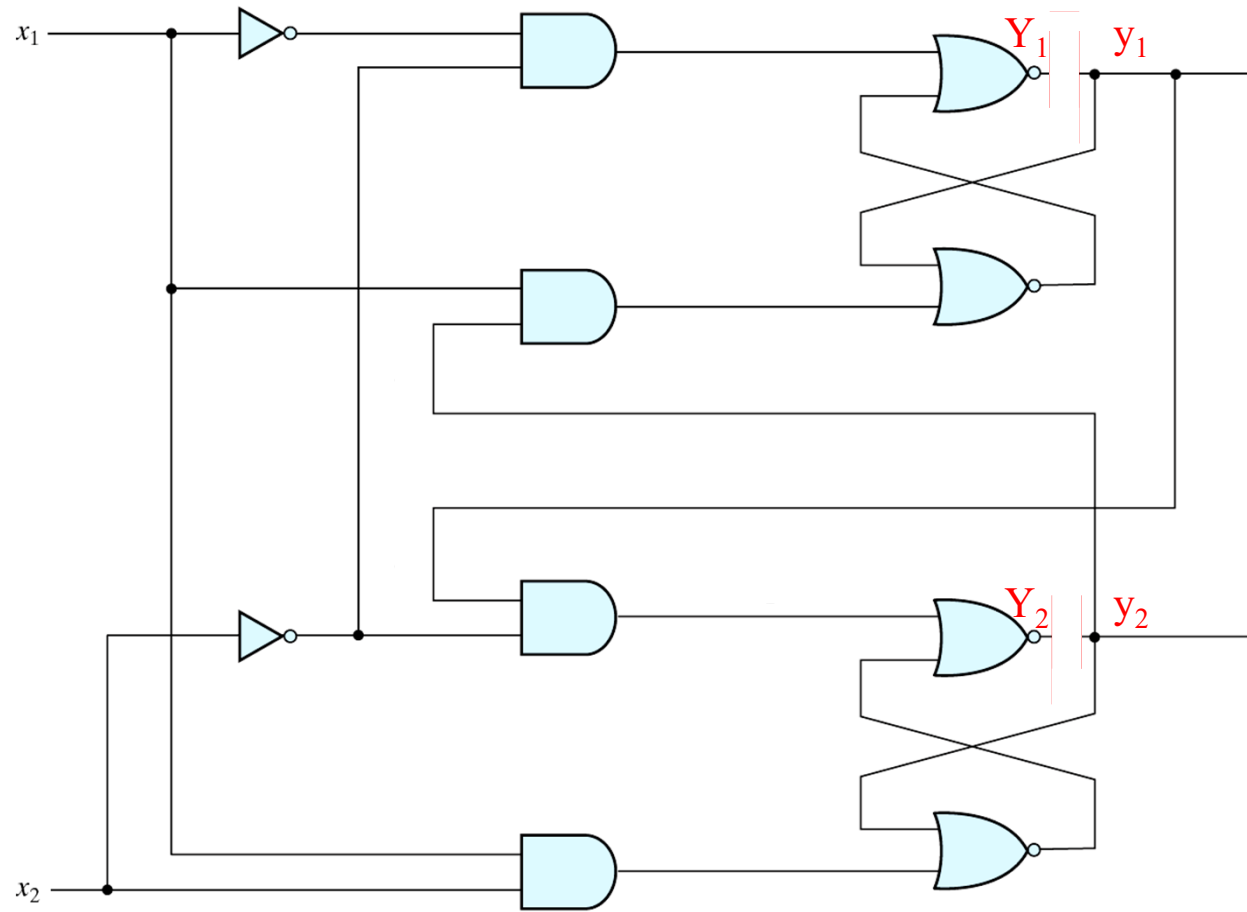
$y \backslash SR$	00	01	11	10
0	1	1	0	0
1	1	1	1	0

(d) Transition table

$S=0, R=0$  ( $S+R=0$ )  
 should not be used  
 $\Rightarrow S+R=1$  is  
 normal mode  
 (eq.  $S'R'=0$ )  
**should be  
 carefully  
 checked first, so it  
 is obtained**  
 $Y = S' + Ry$

# Analysis Example

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## Analysis Example

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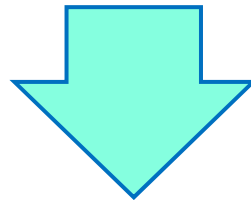
- ✓ The procedure for analyzing an asynchronous sequential circuit with SR latches can be summarized as follows:
  - Label each latch output with  $Y_i$  and its external feedback path with  $y_i$  for  $i=1,2,\dots,k$
  - Derive the Boolean functions for the  $S_i$  and  $R_i$  inputs in each latch.

$$S_1 = x_1 y_2$$

$$S_2 = x_1 x_2$$

$$R_1 = x_1' x_2'$$

$$R_2 = x_2' y_1$$



$$S_1 R_1 = x_1 y_2 x_1' x_2' = 0$$

$$S_2 R_2 = x_1 x_2 x_2' y_1 = 0$$

## Analysis Example

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- Check whether  $SR = 0$  for each NOR latch or whether  $S'R' = 0$  for each NAND latch. (if either of these two conditions is not satisfied, there is a possibility that the circuit may not operate properly)

$$S_1 R_1 = x_1 y_2 x_1' x_2' = 0$$

$$S_2 R_2 = x_1 x_2 x_2' y_1 = 0$$

- Evaluate  $Y = S + R'y$  for each NOR latch or  $Y = S' + Ry$  for each NAND latch.

$$Y_1 = S_1 + R_1' y_1 = x_1 y_2 + x_1 y_1 + x_2 y_1$$

$$Y_2 = S_2 + R_2' y_2 = x_1 x_2 + x_2 y_2 + y_1' y_2$$

# Analysis Example

- Construct a map, with the y's representing the rows and the x inputs representing the columns.
- Plot the value of  $Y=Y_1Y_2\dots Y_k$  in the map.
- Circle all stable states such that  $Y=y$ . The result is then the transition table.
- The transition table shows that the circuit is **stable**
- Race Conditions: there is a **critical race** condition when the circuit is initially in total state  $y_1y_2x_1x_2 = 1101$  and  $x_2$  changes from 1 to 0.
- The circuit should go to the total state 0000.
- If  $Y_1$  changes to 0 before  $Y_2$ , the circuit goes to total state **0100** instead of **0000**.

	$x_1x_2$			
	00	01	11	10
$y_1y_2$ 00	00	00	01	00
01	01	01	11	11
11	00	11	11	10
10	00	10	11	10

Transition Table

$$Y_1 = x_1y_2 + x_1y_1 + x_2y_1$$

$$Y_2 = x_1x_2 + x_2y_2 + y_1'y_2$$

# Implementation Procedure

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- ✓ Procedure to implement an asynchronous sequential circuits with SR latches:
  - Given a transition table that specifies the excitation function  $Y = f(y_1, \dots, y_n, x_1, \dots, x_m)$  derive a pair of maps for each  $S_i$  and  $R_i$  using the latch excitation table
  - Derive the Boolean functions for each  $S_i$  and  $R_i$  (do not to make  $S_i$  and  $R_i$  equal to 1 in the same minterm square; for NAND latch, use the complemented values)
  - Draw the logic diagram using  $k$  latches together with the gates required to generate the  $S$  and  $R$

# Implementation Example

- ✓ Given a transition table  $Y = f(y_1, \dots, y_n, x_1, \dots, x_m)$ , then the general procedure for implementing a circuit with SR latches is specified by the excitation function, and can be summarized as follows:

- Given a transition table

		$x_1x_2$			
	$y$	00	01	11	10
0		0	0	0	1
1		0	0	1	1

(a) Transition table

$$Y = x_1x'_2 + x_1y$$

- Determine the Boolean functions for the S and R inputs of each latch (this is done by using the latch excitation table)

		$x_1x_2$			
	$y$	00	01	11	10
0		0	0	0	1
1		0	0	X	X

(c) Map for  $S = x_1x'_2$

		$x_1x_2$			
	$y$	00	01	11	10
0		X	X	X	0
1		1	1	0	0

(d) Map for  $R = x'_1$

# Implementation Example

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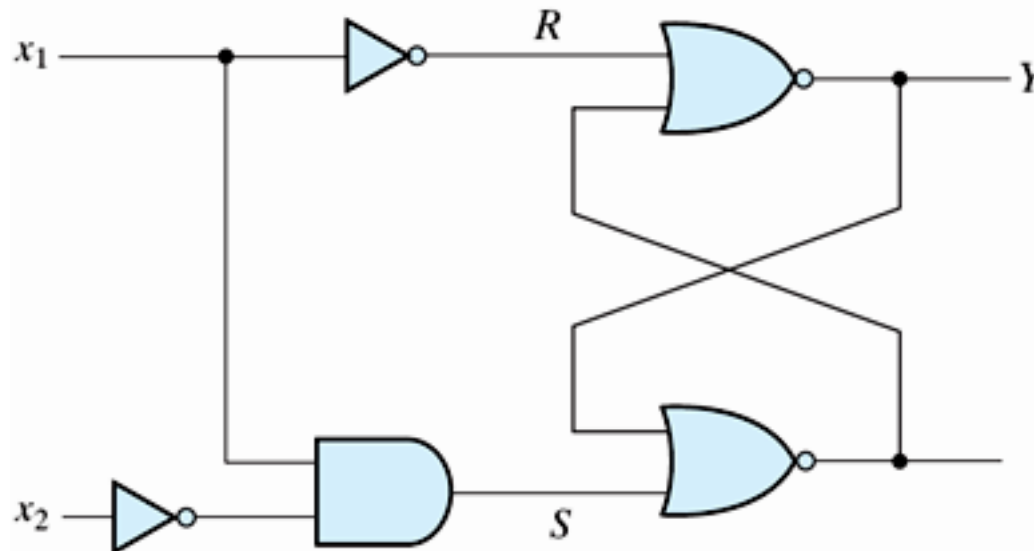
- From maps: the simplified Boolean functions are

$$S = x_1x_2' \quad \text{and} \quad R = x_1' \quad \Rightarrow \quad \text{NOR latch}$$

- Check whether  $SR=0$  for each NOR latch or whether  $S'R'=0$  for each NAND latch:

$$SR = x_1x_2'x_1' = 0$$

- Draw the logic diagram, using k latches together with the gates required to generate the S and R Boolean functions obtained in step1 (for NAND latches, use the complemented values)



# Primitive Flow Table

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- ✓ Primitive flow table - has exactly one stable total state (internal state + input) per row
- ✓ To avoid the timing problems:
  - Only one input variable changes at a time
  - Networks reach a stable total state between input changes (Fundamental Mode)
- ✓ Every change in input changes the state

## Design procedure

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1. Obtain a **primitive table** from specifications
2. Reduce flow table by merging rows in the primitive flow table
3. **Assign binary state variables** to each row of reduced table
4. **Assign output values** to dashes associated with unstable states to obtain the output map
5. **Simplify Boolean** functions for excitation and output variables;
6. **Draw the logic diagram**



# Design Example 1:

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## ✓ Problem Statement:

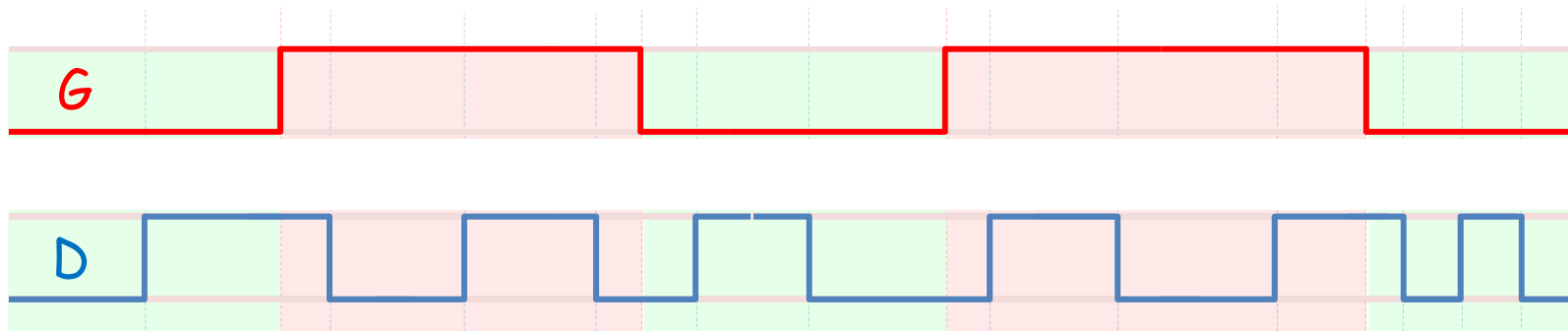
- Design a gated latch circuit (memory element) with two inputs,  $G$ (gate) and  $D$ (Data) and one output  $Q$ .
- The  $Q$  output will follow the  $D$  input as long as  $G=1$ . When  $G$  goes to 0, the information that was present at the  $D$  input at the time of transition is retained at the  $Q$  output.
  - $Q = D$  when  $G = 1$
  - $Q$  retains its value when  $G$  goes to 0

# Design Example 1:

## 1-Primitive Flow Table

- ✓ A primitive flow table is a flow table with only one stable total state (internal state + input) in each row.
- ✓ In order to form the primitive flow table, we first form a table with all possible total states, combinations of the inputs and internal states, simultaneous transitions of two input variables are not allowed

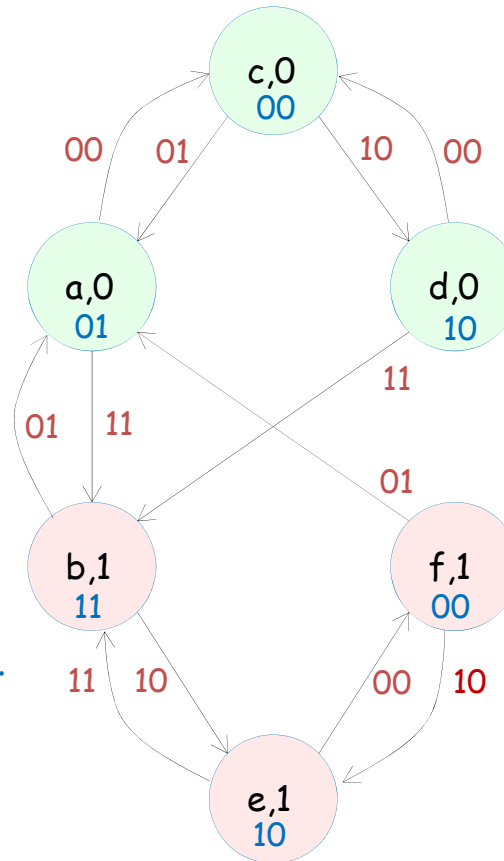
State	Inputs		Output	Comments
	D	G	Q	
<i>a</i>	0	1	0	$D = Q$ because $G = 1$
<i>b</i>	1	1	1	$D = Q$ because $G = 1$
<i>c</i>	0	0	0	After state <i>a</i> or <i>d</i>
<i>d</i>	1	0	0	After state <i>c</i>
<i>e</i>	1	0	1	After state <i>b</i> or <i>f</i>
<i>f</i>	0	0	1	After state <i>e</i>



# Design Example 1

## 1-Primitive Flow Table

- ✓ One square in each row is a stable state for that row.
- ✓ First, we note that both inputs are not allowed to change at the same time.
  - We enter dash marks in each row that differs in two or more variables from the input variables associated with the stable state.
- ✓ Next it is necessary to find values for the two squares adjacent to the stable state in each row.
  - the previous table may support in deriving the necessary information.
- ✓ All outputs associated with unstable states are don't care conditions
  - We marked them with a dash.



State	Inputs		Output	Comments
	D	G	Q	
a	0	1	0	D = Q because G = 1
b	1	1	1	D = Q because G = 1
c	0	0	0	After state a or d
d	1	0	0	After state c
e	1	0	1	After state b or f
f	0	0	1	After state e

	DG			
	00	01	11	10
a	c, -	a, 0	b, -	-, -
b	-, -	a, -	b, 1	e, -
c	c, 0	a, -	-, -	d, -
d	c, -	-, -	b, -	d, 0
e	f, -	-, -	b, -	e, 1
f	f, 1	a, -	-, -	e, -

# Design Example 1

## 2-Reduction of the Primitive Flow Table

- Two or more rows can be merged into one row if there are non-conflicting states and outputs in every columns.

Candidates states for merging:

DG		DG			
		00	01	11	10
States	a	c, -	a, 0	b, -	-, -
	c	c, 0	a, -	-, -	d, -
	d	c, -	-, -	b, -	d, 0

DG		DG			
		00	01	11	10
States	b	-, -	a, -	b, 1	e, -
	e	f, -	-, -	b, -	e, 1
	f	f, 1	a, -	-, -	e, -

- After merged into one row:
  - Don't care entries are overwritten
  - Stable states and output values are included
  - A common symbol is given to the merged row

DG		DG			
		00	01	11	10
States	a, c, d	c, 0	a, 0	b, -	d, 0
	b, e, f	f, 1	a, -	b, 1	e, 1

DG		DG			
		00	01	11	10
States	a	a, 0	a, 0	b, -	a, 0
	b	b, 1	a, -	b, 1	b, 1

# Design Example 1

		<i>DG</i>			
		00	01	11	10
States	<i>a</i>	<i>a</i> , 0	<i>a</i> , 0	<i>b</i> , -	<i>a</i> , 0
	<i>b</i>	<i>b</i> , 1	<i>a</i> , -	<i>b</i> , 1	<i>b</i> , 1

## 3-Transition Table and Logic Diagram

- ✓ In order to obtain the circuit described by the reduced flow table, it is necessary to assign a distinct binary value to each state.
- ✓ This converts the flow table to a transition table.
- ✓ A binary state assignment must be made to ensure that the circuit will be free of critical race.

**a=0, b=1 in this example**

Transition table and output map

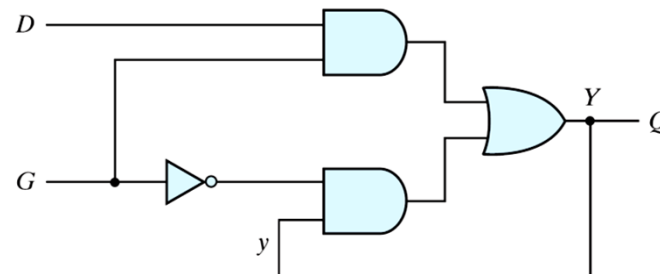
<i>y</i> \ <i>DG</i>		00	01	11	10
		0	0	1	0
<i>y</i>	1	1	0	1	1

(a)  $Y = DG + G'y$

<i>y</i> \ <i>DG</i>		00	01	11	10
		0	0	0	0
<i>y</i>	1	1	1	1	1

(b)  $Q = Y$

Gated-latch logic diagram



# Design Example 1

## 4. Implementation with SR Latch

	<i>DG</i>			
<i>y</i>	00	01	11	10
0	0	0	1	0
1	1	0	1	1

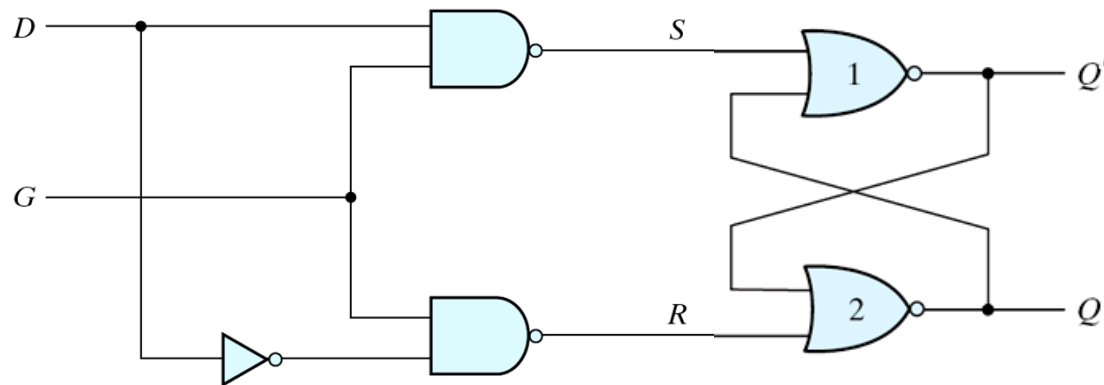
	<i>DG</i>			
<i>y</i>	00	01	11	10
0	0	0	1	0
1	X	0	X	X

(a)  $S = DG$

	<i>DG</i>			
<i>y</i>	00	01	11	10
0	X	X	0	X
1	0	1	0	0

$R = D'G$

Circuit with SR latch



(b) Logic diagram

# Design Example:

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## 5- Assigning Outputs to Unstable States

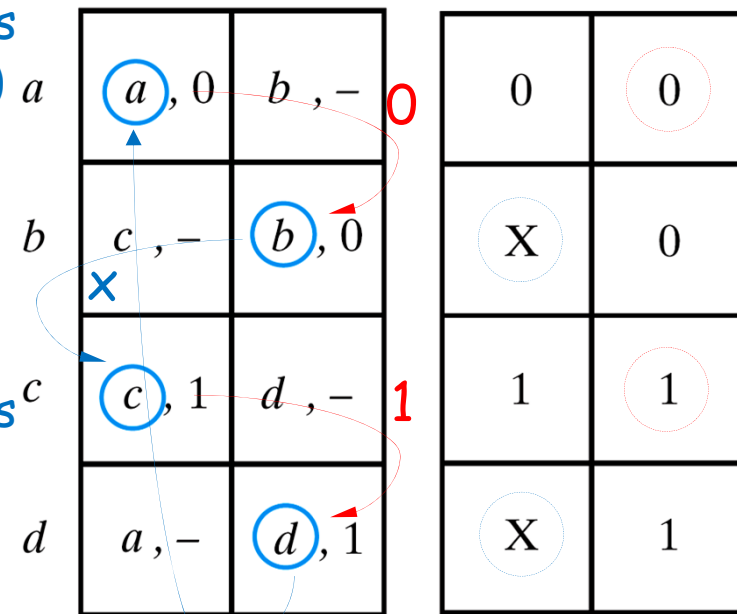
- ✓ While the stable states in a flow table have specific output values associated with them, the unstable states have unspecified output entries designated by a dash.
- ✓ These unspecified output values must be chosen so that **no momentary false outputs** occur when the circuit switches between stable states.
  - *If the two stable states have the same output value, then an unstable states that are a transient state between them must have the same output.*
  - *If an output variable is to change as a result of a state change, then this variable is assigned a don't care condition\*.*

# Design Example 1

## 5- Assigning Outputs to Unstable States

Example:

- If a changes to b, the two stable states have the same output value = 0 ( $0 \Rightarrow 0$ : 0) the transient unstable state b in the first row must have the same output value = 0
- if c changes to d same for  $1 \Rightarrow 1$ : 1
- If b changes to c, the two stable states have different output values  $0 \Rightarrow 1$ : x the transient unstable state c in the second row is assigned a don't care condition
- if d changes to a same for  $1 \Rightarrow 0$ : x



(a) Flow table b) Output assignment